OKI Semiconductor MSC1157

Speaker Drive Amplifier

GENERAL DESCRIPTION

The MSC1157, designed specifically to operate at a low voltage with low current consumption, is a power amplifier developed for driving a speaker for a voice IC.

The voltage gains can be adjusted over a range of up to ten. The differential output can directly drive a speaker without any output coupling capacitors. The MSC 1157, because of its ability to stand by, is ideally suitable for portable equipment applications powered by a battery.

FEATURES

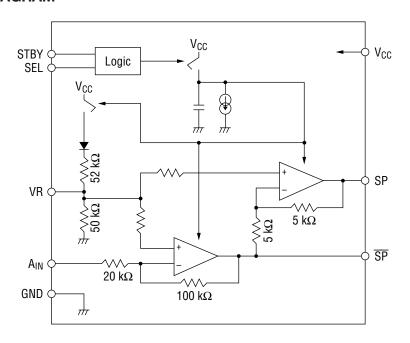
- Low voltage operation
- Low current dissipation Operating current
- Standby function
- High output current
- Differential outputs
- Adjustable gain

- : 2.0 to 6.0 V (Single power supply)
- : 1.6mA without load (typ.)
- : Current dissipation less than 1 μA in standby

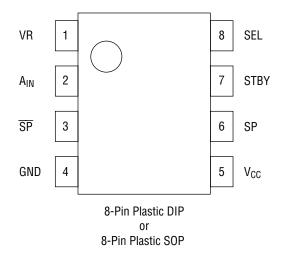
: 350mA peak

- : A speaker can be directly connected between differential outputs.
- : Gain can be adjusted by use of an external resistor.
- Package options:
 8-pin plastic DIP (DIP8-P-300-2.54) (Product name : MSC1157RS)
 8-pin plastic SOP (SOP8-P-250-1.27-K) (Product name : MSC1157MS-K) Chip

BLOCK DIAGRAM



PIN CONFIGURATION (TOP VIEW)



PIN DESCRIPTIONS

Pin	Symbol	Туре	Description						
5	V _{CC}	_	Power supply pin.						
4	GND	_	Ground pin.						
2	A _{IN}	I	Signal input pin for analog signal inputs, etc.						
7, 8	STBY, SEL	Ι	Digital input pins. Setting these pins for how to set the pins. Applying a clock between 32kHz and to operation status regardless of the of the pins at the same time may can <u>Refer to the section, RECOMMENDE</u> are changed by setting the SEL pin.	SEL 0 1 Clock 4 4MHz to either status set at the use malfunction.	STBY 0 1 Clock 0 1 Clock 0 1 Clock the STBY or th e other pin. Ap	StatusOperationStandbyOperationStandbyOperationOperationOperationOperationOperationUnstable Operationunstable Operatione SEL pin leads the ICplying clocks to both			
1	VR	0	Bias output pin for internal circuits. This pin is at GND potential during standby. Connecting a capacitor between VR and the GND pin reduces the pop-up noise at power on and improves the ripple elimination ratio.						
3	SP	0	Speaker output pin. This pin outputs a negative phase with respect to the input signal.						
6	SP	0	Speaker output pin. This pin outputs a positive phase with respect to the input signal.						

Parameter	Symbol	Condition	Rating	Unit	Remark
Power Supply Voltage	V _{CC}	Ta=25°C	-0.3 to +6.5	V	V _{CC}
Input Voltage	V	Ta=25°C	-0.3 to V _{CC} +0.3	V	STBY
Input Voltage	VIN				A _{IN} , SEL
Mavimum Output Current		Ta=25°C	(*1)		
Maximum Output Current	IOMAX		±400	mA	SP, SP
Dower Discinction	п		470	mW	DIP type
Power Dissipation	PD	Ta=25°C	400	mW	SOP type
Junction Temperature	T _{jMAX}	_	125	°C	Chip
Storage Temperature	T _{STG}	_	-55 to +150	°C	

ABSOLUTE MAXIMUM RATINGS

*1 Avoid shorting the output pins (SP and \overline{SP}) to V_{CC} or GND because the IC may be damaged.

RECOMMENDED OPERATING CONDITIONS

Parameter	Symbol	Condition	Min.	Max.	Unit
Power Supply Voltage	V _{CC}	—	2.0	6.0	V
Load Impedance (*2)	RL	—	8.0	_	Ω
Peak Load Current	I _{0-Р}	—	—	350	mA
"H" Input Voltage	VIH	For CTDV and CEL pipe	0.7 V _{CC}	—	V
"L" Input Voltage	VIL	For STBY and SEL pins	—	0.3 V _{CC}	V
	fstby	SEL = "L"		4.096 M	- Hz
		At clock input	32 k		
CTDV Operating Frequency (*2)		$V_{CC} \ge 2.4 V$			
STBY Operating Frequency (*3)		SEL = "H"		1 M	
		At clock input	32 k		
		$V_{CC} \ge 2.4 V$			
Operating Temperature	Тор	_	-20	+70	°C

*2 A speaker of 8 Ω (standard) or more should be used.

*3 The input of clocks may cause a little noise in output waveforms.

It is recommended to input the DC voltage to inprove voice quality.

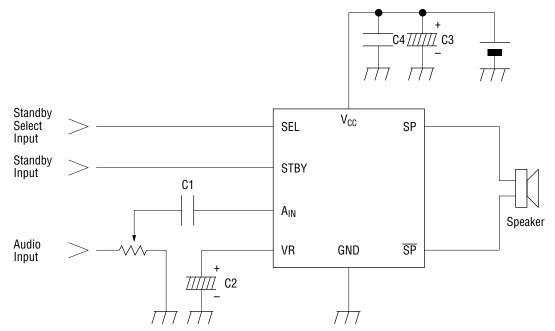
Parameter	Symbol	Condition		Min.	Тур.	Max.	Unit
AIN Input Resistance	RIN			14	20	26	kΩ
	A _{V1}	v ₂ <u>SP</u> →SP		13.44	14	14.49	
Voltage Gain	A _{V2}			-1.94	0	+1.58	dB
	A _{V3}			19.46	20	20.51	
Output Douge	P _{OUT1}	V _{CC} =3 V, f=1 kHz RL=8 Ω, THD≥10%		100	178	_	mW
Output Power	P _{OUT2}	V _{CC} =6 V, f=1 kHz RL=32 Ω, THD≥10%		300	440		mW
Total Harmonia Distantian	THD1	V _{CC} =3 V, RL=8 Ω f=1 kHz, P _{OUT} =45 mW			1.2	_	%
Total Harmonic Distortion	THD2	V _{CC} =6 V, RL=32 Ω f=1 kHz, P _{OUT} =125 mW		_	0.37	_	%
Ripple Elimination Ratio	RR	f=1 kHz, C2=4.7 μF		30	43	_	dB
Output DC Voltage	V ₀	In no	V _{CC} =2 V	0.53	0.65	0.77	- V
(*4)		signal state	V _{CC} =6 V	2.49	2.61	2.73	
Output Offset Voltage	ΔV_0	Betwee	en SP- <u>SP</u>			±30	mV
Output "H" Voltage	ut "H" Voltage V _{OH} A _{IN} =V _{CC} or GND I _{OUT} =-100 mA		-	V _{CC} -1.15	V _{CC} -1.04	—	V
Output "L" Voltage	V _{OL}	A _{IN} =V _{CC} or GND I _{OUT} =100 mA		_	0.17	0.3	V
STBY, SEL	IIH	V _I =V _{CC}			—	±0.1	μA
Input Current	Ι _{ΙL}	V _I =GND			—	±0.1	μA
VR Equivalent Resistance	R _{VR}	_		18	25	32	kΩ
Circuit Current During Operation	I _{CC}	V _{CC} =6 V, RL=∞		1.1	1.6	2.4	mA
Circuit Current During Standby	I _{CCS}	_			_	1.0	μA

ELECTRICAL CHARACTERISTICS

*4 The typical value of the output voltage in no signal state is determined from the following equation.

$$V_{\rm O} = (V_{\rm CC} - 0.67) - \frac{50 \text{ k}\Omega}{50 \text{ k}\Omega + 52 \text{ k}\Omega}$$

APPLICATION CIRCUIT



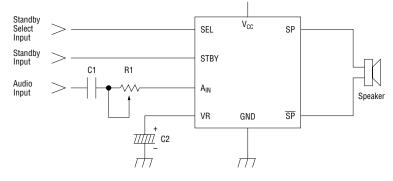
- If parasitic capacitance of 60pF or more exists between GND and the speaker output pin SP or SP, oscillation may occur. Implement the circuit mount design so as to be less than 60pF.
- C1 is the AC coupling capacitor. Cutoff frequency fc on the low frequency side is determined by the following equation. Choose a value of C1 according to the bandwidth.

$$fc = \frac{1}{2 \times \pi \times C1 \times 20k}$$
(Hz)

- Choose a value of C2 that is 80 to 100 times as large as that of C1.
- When the standby function is not used, connect the pins STBY and SEL to V_{CC} or GND.
- It is recommended that the capacitor C4 (approximately 0.1μ F) having better high frequency characteristics and the capacitor C3 (approximately 10μ F) be placed between the pins V_{CC} and GND.

GAIN ADJUSTMENT

1. Gain Adjustment Using Input Resistance (This approach allows gain adjustment with fewer external components)



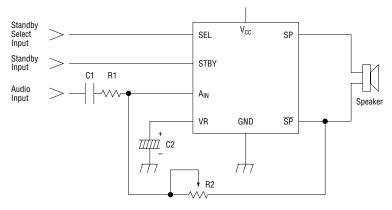
• Cutoff frequency fc on the low frequency side is determined from the equation:

$$fc \doteq \frac{1}{2 \times \pi \times C1 \times (R1 + 20k)}$$
 (Hz)

• Voltage gain A_{V1} is determined from the equation:

$$A_{V1} \doteq \frac{100k}{R1 + 20k} (V/V)$$

2. Gain Adjustment Using Feedback Resistance (This approach has the advantage over the above approach (less noise approach), but the number of components is increased)



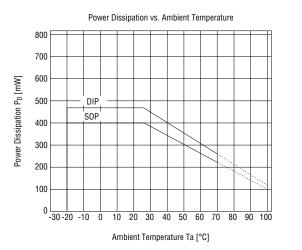
• Cutoff frequency fc on the low frequency side is determined from the equation:

$$fc = \frac{1}{2 \times \pi \times C1 \times Zin}$$
 (Hz) $Zin = R1 + \frac{R2 \times 20k}{R2 + 120k}$ (Ω)

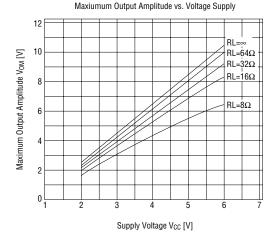
• Voltage gain A_{V1} is determined from the equation:

$$A_{V1} \doteq \frac{5}{1 + \frac{R1}{20k} + \frac{6 \times R1}{R2}} \quad (V/V)$$

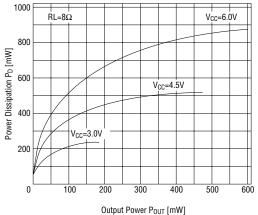
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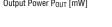


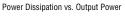
OPERATING CHARACTERISTICS

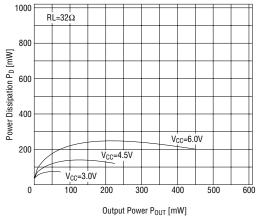


Power Dissipation vs. Output Power

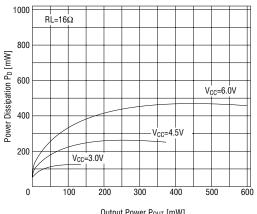




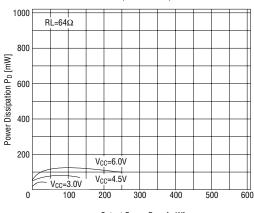




Power Dissipation vs. Output Power



Output Power POUT [mW]



Power Dissipation vs. Output Power

Output Power POUT [mW]

2E-3

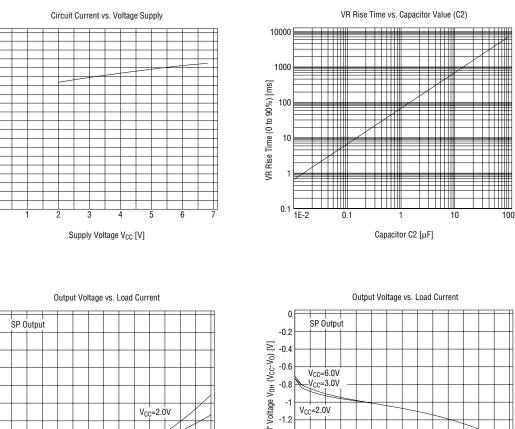
1.5E-3

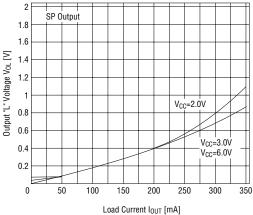
1E-3

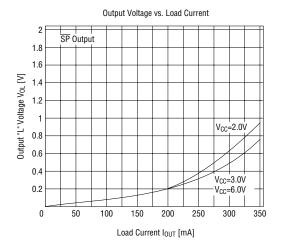
5E-4

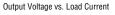
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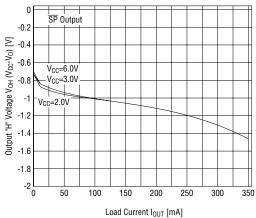
Circuit Current Icc [A]

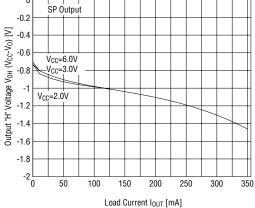




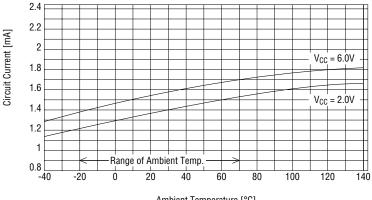




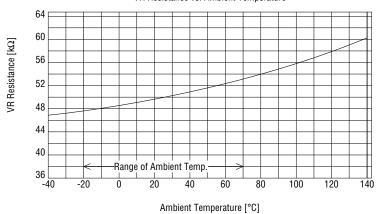


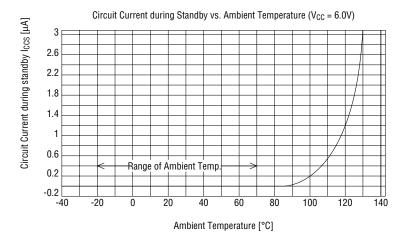




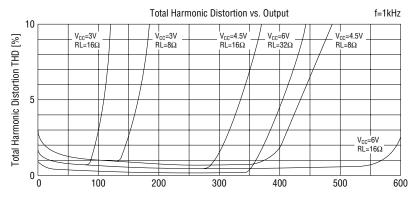


Ambient Temperature [°C]

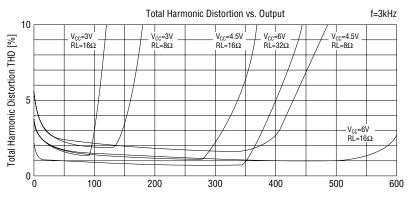




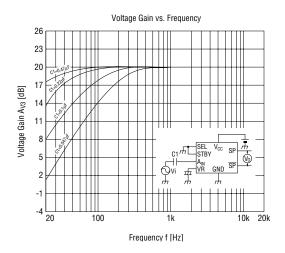
VR Resistance vs. Ambient Temperature



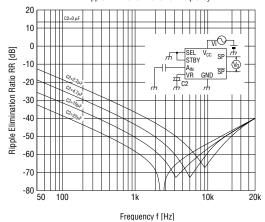
Output Power POUT [mW]



Output Power POUT [mW]



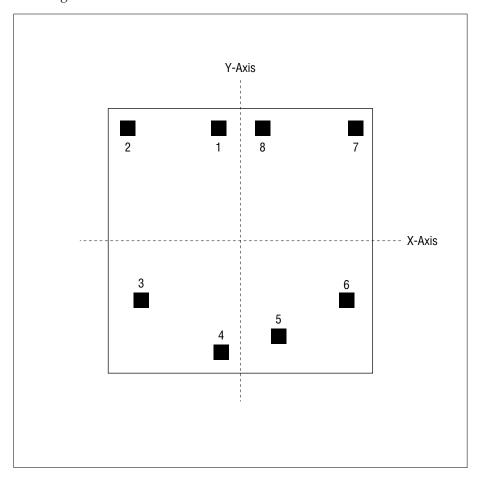
Ripple Elimination Ratio vs. Frequency



PAD CONFIGURATION

Pad Layout

:X=2.3mm, Y=2.4mm
:350±30µm
:110×110µm
: GND



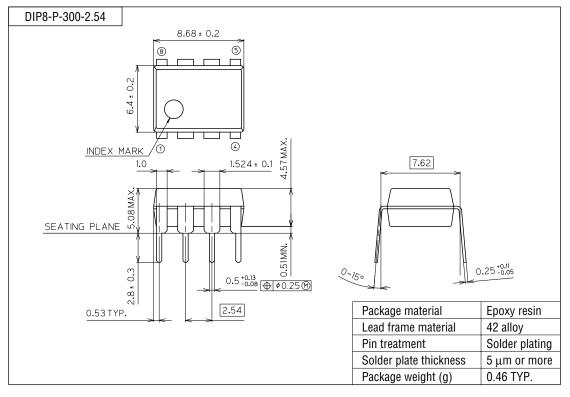
Pad Coordinates

(Chip center is located at X=0 and Y=0.)

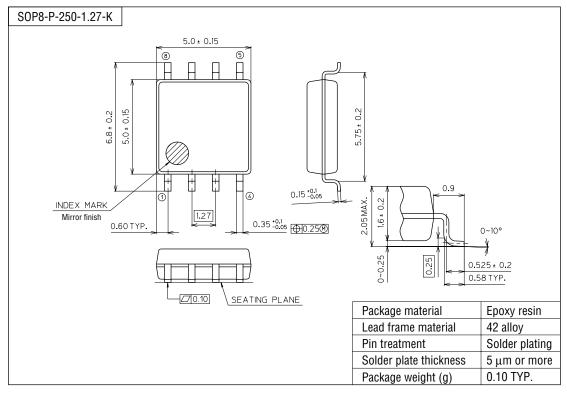
(Chip tenter is located at	X=0 dild 1=0.)		(Unit: μm)
Pad No.	Pad Name	X-AXIS	Y-AXIS
1	VR	-133	1035
2	A _{IN}	-985	1035
3	SP	-950	-263
4	GND	-180	-1027
5	V _{CC}	240	-914
6	SP	950	-263
7	STBY	985	1035
8	SEL	159	1035

PACKAGE DIMENSIONS

(Unit : mm)







Notes for Mounting the Surface Mount Type Package

The SOP, QFP, TSOP, TQFP, LQFP, SOJ, QFJ (PLCC), SHP, and BGA are surface mount type packages, which are very susceptible to heat in reflow mounting and humidity absorbed in storage. Therefore, before you perform reflow mounting, contact Oki's responsible sales person on the product name, package name, pin number, package code and desired mounting conditions (reflow method, temperature and times).

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- 2. The outline of action and examples for application circuits described herein have been chosen as an explanation for the standard action and performance of the product. When planning to use the product, please ensure that the external conditions are reflected in the actual circuit, assembly, and program designs.
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